

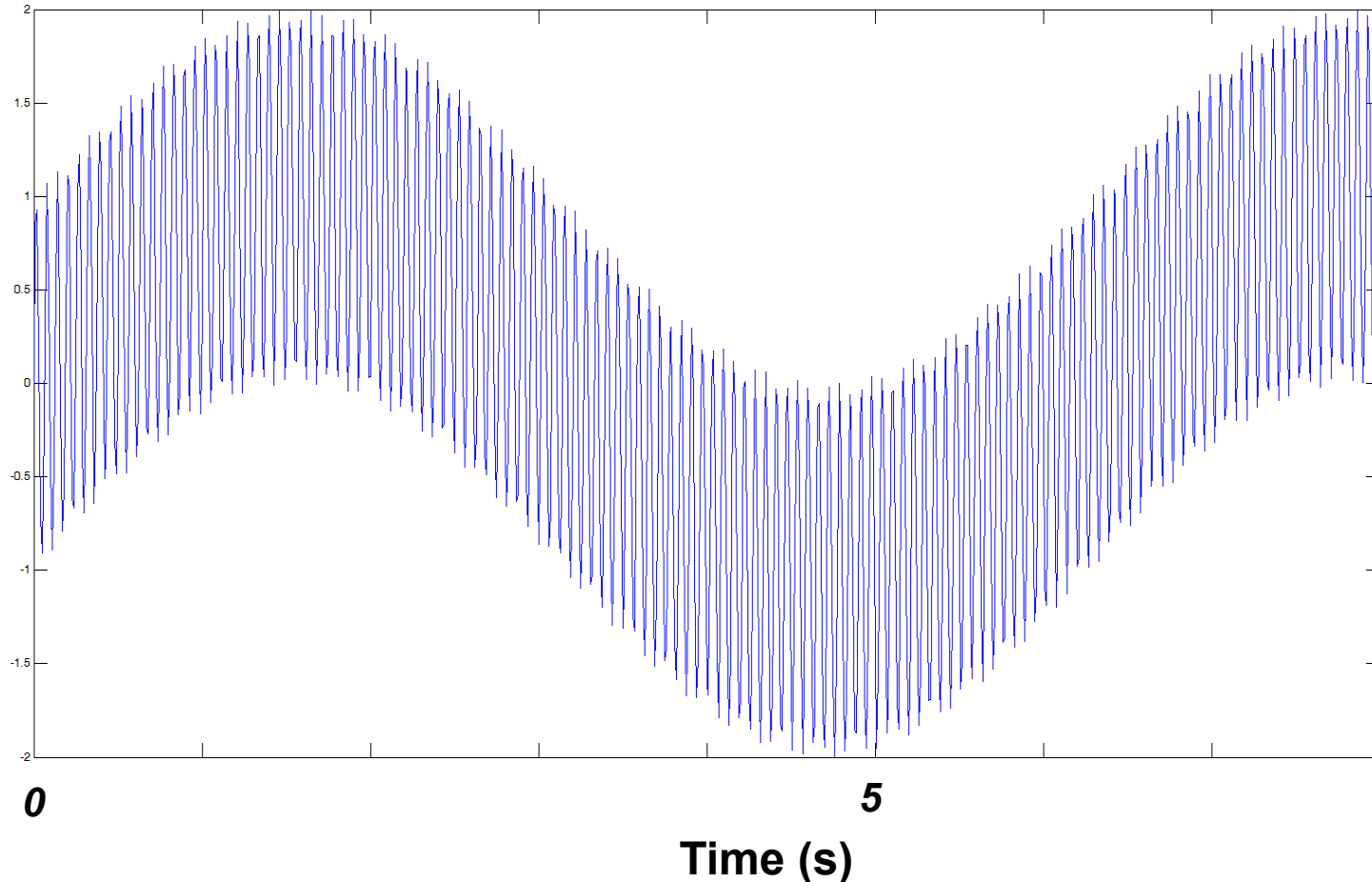
Analog Filtering



COLUMBIA | ENGINEERING
The Fu Foundation School of Engineering and Applied Science



Why do we need filters?



- Filters are used to modify the measured signal by removing unwanted signals, i.e. noise.

Noise

- Any unwanted signal that corrupts the signal of interest.
- Common types of noise are:
 - Extraneous noise
 - Thermal noise
 - power noise
 - $1/f$ noise

Extraneous noise

- Transducers will frequently respond to more than one source.
- Which source is regarded as the signal and which source(s) is regarded as noise is dependent upon the application.
- For instance, electrodes placed on the chest will record the electrical activity of both the heart (ECG) and the respiratory muscles (EMG).

Thermal noise

- In conductors of non-zero resistance, thermally excited electron currents flow randomly and sum statistically to produce a noise voltage across the conductor's end terminals. This effect is called thermal noise. (Also called Johnson-Nyquist noise).
- At frequencies below 100 MHz, the root mean square (rms) amplitude of thermal noise for a resistor is given by Nyquist's relation:

$$v_{rms} = \sqrt{4KTR\Delta f}$$

Where K is Boltzmann's constant (1.38×10^{-23} joules/K), T is absolute temperature, R is resistance, and Δf is system bandwidth

Thermal noise

- For a neural recording electrode, its impedance is 5M Ω , and the system bandwidth is 20KHz. Thus the thermal noise is

$$v_{rms} = \sqrt{4KTR\Delta f} \approx 40\mu V$$

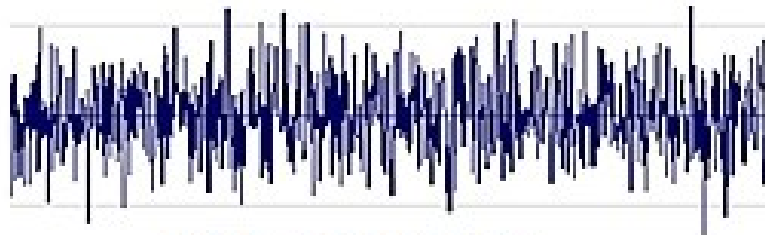
Power noise

- Capacitive and/or inductive coupling from AC power can induce 60 Hz noise in wiring and instrumentation.
- Although, power noise has a basic frequency of 60 Hz, nonlinear coupling may give rise to other components at higher (or lower) frequencies.
 - Usually these will occur at integer multiples ($60N$) or sub-multiples ($60/N$) of 60Hz.
- Always be suspicious of signals with strong components at such frequencies (i.e. 15, 30, 60, 120, 180 Hz)

1/f noise (pink noise)

- Caused by fluctuation of resistance of materials
- In contrast to white noise, 1/f noise exhibits a pattern that high frequency noise rides on larger low frequency content.
- Noise amplitude is inversely proportional to the frequency

Flat-band or
White Noise

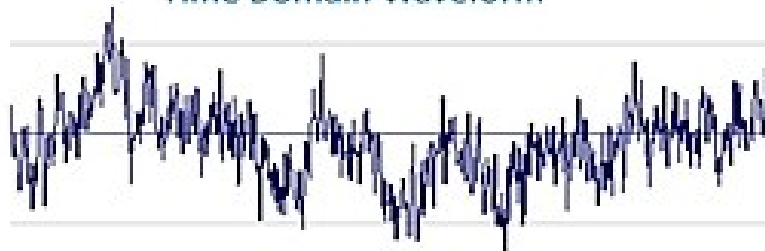


Time Domain Waveform



Spectral Density

1/f flicker
noise

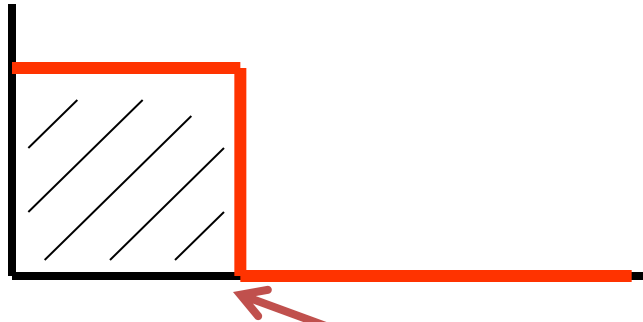


Analog filters

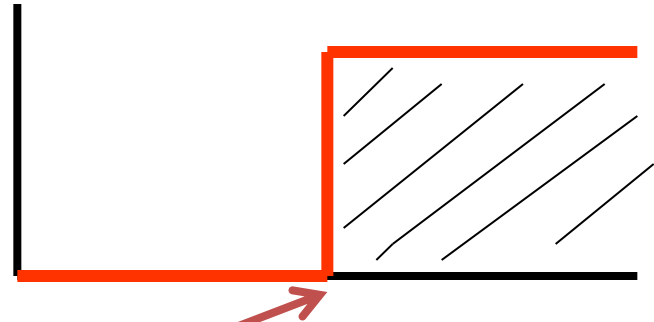
- Are usually implemented with R, L, and C components and operational amplifiers.
- May be classified as either passive or active
 - A passive filter typically contains only R, L, and C components. It is not necessary that all three be present.
 - An active filter contains, along with R, L, and C components, an energy source, such as an operational amplifier.

Four types of filters

Low-pass

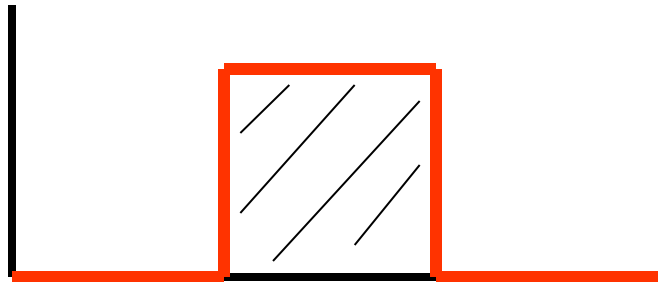


High-pass

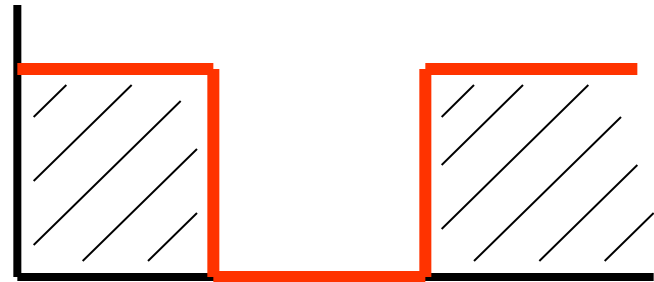


Cut-off frequency

Band-pass



Band-stop



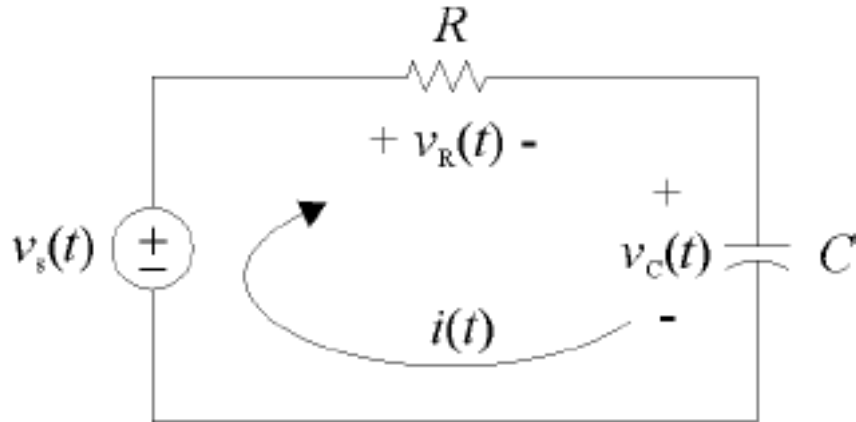
Filter characteristics

- The *cutoff frequency* or *corner frequency* of a filter is used to describe the transition point from the pass band to the reject band.
- Since this cannot occur instantaneously, it is usually defined as the point where the filter output has decreased to 0.707 of the value in the pass band.
- This corresponds to a gain of -3dB and so the cutoff frequency is sometimes referred to as the - 3dB point.

Filter characteristics

- The *roll-off* refers to rate at which the filter attenuates the input after the cutoff point.
- In many cases, the filter gain will decrease linearly on the Bode plot so that the roll off can be specified as the linear slope.
- Thus roll-off may be quoted in terms of dB/decade where a decade refers to a factor of ten in frequency.
- Sometimes roll-off will be quoted in terms of dB/octave where an octave refers to a doubling of frequency.

Low pass filter



- Input signal:

$$v_s(t) = \sin(t) + \sin(100t)$$


- What is the voltage across the capacitor?

$$V_c(s) = \frac{1}{1 + RCs} V_s(s)$$


$$V_s(s) = \frac{1}{s^2 + 1} + \frac{100}{s^2 + 100^2}$$

Low-pass filter

$$\begin{aligned} V_c(s) &= \frac{1}{1 + RCs} \cdot \left(\frac{1}{s^2 + 1} + \frac{100}{s^2 + 100^2} \right) \\ &= \frac{1}{1 + RCs} \cdot \frac{1}{s^2 + 1} + \frac{1}{1 + RCs} \cdot \frac{100}{s^2 + 100^2} \end{aligned}$$



Response to sinusoidal input with a
freq. of 1 rad/s



Response to sinusoidal input with a
freq. of 100 rad/s

Low-pass filter

- Response to the 1 rad/s sinusoidal signal

$$V_{c1}(s) = \frac{1/RC}{s + 1/RC} \cdot \frac{1}{s^2 + 1}$$

$$L^{-1}[V_{c1}(s)] = \frac{1}{RC} \left(\frac{1}{(\frac{1}{RC})^2 + 1} e^{-\frac{t}{RC}} + \frac{1}{\sqrt{(\frac{1}{RC})^2 + 1}} \sin(t - \theta) \right)$$

$$= \frac{RC}{(RC)^2 + 1} e^{-\frac{t}{RC}} + \frac{1}{\sqrt{(RC)^2 + 1}} \sin(t - \theta)$$

Transient response

Steady-state response

Where: $\theta = \tan^{-1}(RC)$

Low-pass filter

- Response to the 100 rad/s sinusoidal signal

$$V_{c2}(s) = \frac{\frac{1}{RC}}{s + \frac{1}{RC}} \cdot \frac{100}{s^2 + 100^2}$$

$$\begin{aligned} L^{-1}[V_{c2}(s)] &= \frac{1}{RC} \left(\frac{100}{\left(\frac{1}{RC}\right)^2 + 100^2} e^{-\frac{t}{RC}} + \frac{1}{\sqrt{\left(\frac{1}{RC}\right)^2 + 100^2}} \sin(100t - \theta) \right) \\ &= \frac{100RC}{(100RC)^2 + 1} e^{-\frac{t}{RC}} + \frac{1}{\sqrt{(100RC)^2 + 1}} \sin(100t - \theta) \end{aligned}$$

$$\text{Where: } \theta = \tan^{-1}(100RC)$$

Low-pass filter

$$v_s(t) = \sin(t) + \sin(100t)$$

- If $RC = 1$

$$V_{c1}(t) = \frac{1}{\sqrt{(RC)^2 + 1}} \sin(t - \theta) = 0.707 \cdot \sin(t - \theta)$$

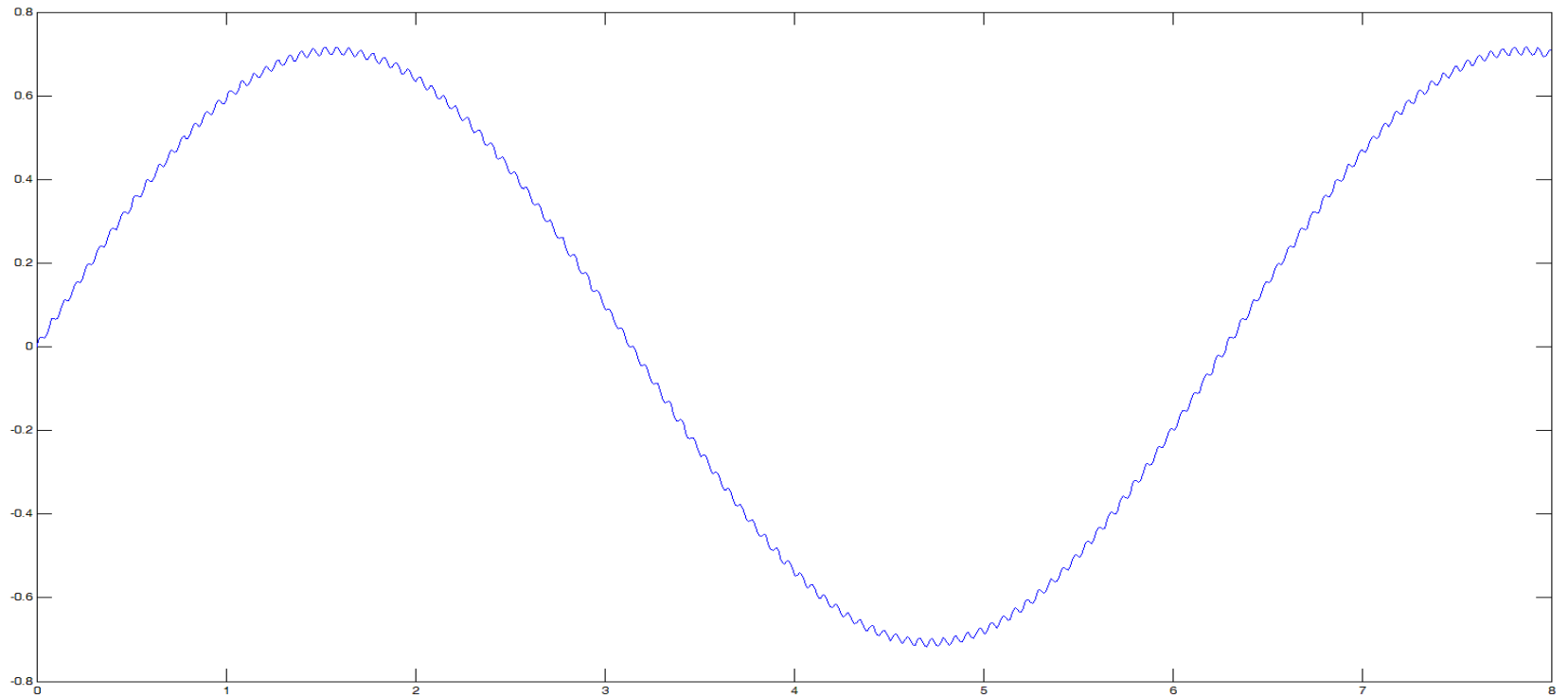
$$V_{c2}(t) = \frac{1}{\sqrt{(100RC)^2 + 1}} \sin(100t - \theta)$$

$$= \frac{1}{\sqrt{10001}} \sin(100t - \theta) \approx 0.01 \cdot \sin(100t - \theta)$$

High frequency signal was attenuated by 99% while the low frequency signal was attenuated by 30%

Low-pass filter

Filtered signal



Low-pass filter

- What if $RC = 0.001$?

$$V_{c1}(t) = \frac{1}{\sqrt{(RC)^2 + 1}} \sin(t - \theta) \approx \sin(t - \theta)$$

$$V_{c2}(t) = \frac{1}{\sqrt{(100RC)^2 + 1}} \sin(100t - \theta) \approx \sin(100t - \theta)$$

- Neither the high frequency signal nor the low frequency signal was attenuated!
- RC determines the cut-off frequency of the filter.

Low-pass filter

- What if $RC = 100$?

$$V_{c1}(t) = \frac{1}{\sqrt{(RC)^2 + 1}} \sin(t - \theta) \approx 0.01 \cdot \sin(t - \theta)$$

$$V_{c2}(t) = \frac{1}{\sqrt{(100RC)^2 + 1}} \sin(100t - \theta) \approx 0.0001 \sin(100t - \theta)$$

- Both the high frequency signal and the low frequency signal was attenuated.

Frequency response of a RC circuit

- Response to a sinusoidal signal with angular frequency ω

$$V_c(s) = \frac{\frac{1}{RC}}{s + \frac{1}{RC}} \cdot \frac{\omega}{s^2 + \omega^2}$$

$$\begin{aligned} L^{-1}[V_c(s)] &= \frac{1}{RC} \left(\frac{\omega}{\left(\frac{1}{RC}\right)^2 + \omega^2} e^{-\frac{t}{RC}} + \frac{1}{\sqrt{\left(\frac{1}{RC}\right)^2 + \omega^2}} \sin(\omega t - \theta) \right) \\ &= \frac{\omega RC}{(\omega RC)^2 + 1} e^{-\frac{t}{RC}} + \frac{1}{\sqrt{(\omega RC)^2 + 1}} \sin(\omega t - \theta) \end{aligned}$$

$$\text{Where: } \theta = \tan^{-1}(\omega RC)$$

Frequency response of a RC circuit

- Steady state response

$$V_c(t) = \frac{1}{\sqrt{(\omega RC)^2 + 1}} \sin(\omega t - \theta)$$

Where: $\theta = \tan^{-1}(\omega RC)$

- If $RC = 1$

$$\omega = 0.01, \quad \frac{1}{\sqrt{(\omega RC)^2 + 1}} \approx 1$$

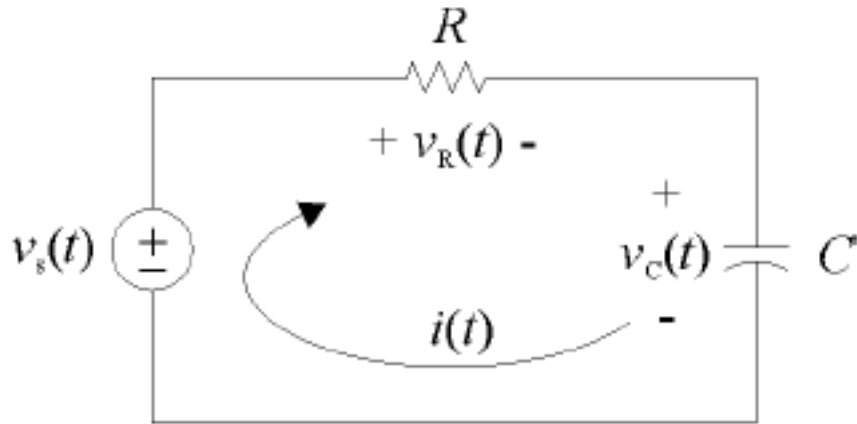
$$\omega = 0.1, \quad \frac{1}{\sqrt{(\omega RC)^2 + 1}} \approx 0.995$$

$$\omega = 1, \quad \frac{1}{\sqrt{(\omega RC)^2 + 1}} \approx 0.707$$

$$\omega = 10, \quad \frac{1}{\sqrt{(\omega RC)^2 + 1}} \approx 0.1$$

$$\omega = 100, \quad \frac{1}{\sqrt{(\omega RC)^2 + 1}} \approx 0.01$$

High pass filter



- What if I want to preserve the high frequency signal?
- The voltage across the resistor response to a sinusoidal signal $\sin(\omega t)$

$$V_R(s) = \frac{RCs}{1 + RCs} V_s(s)$$

$$\begin{aligned} V_R(s) &= \frac{RCs}{1 + RCs} \cdot \frac{\omega}{s^2 + \omega^2} = \left(1 - \frac{1}{1 + RCs}\right) \cdot \frac{\omega}{s^2 + \omega^2} \\ &= \frac{\omega}{s^2 + \omega^2} - \frac{1}{1 + RCs} \cdot \frac{\omega}{s^2 + \omega^2} \end{aligned}$$